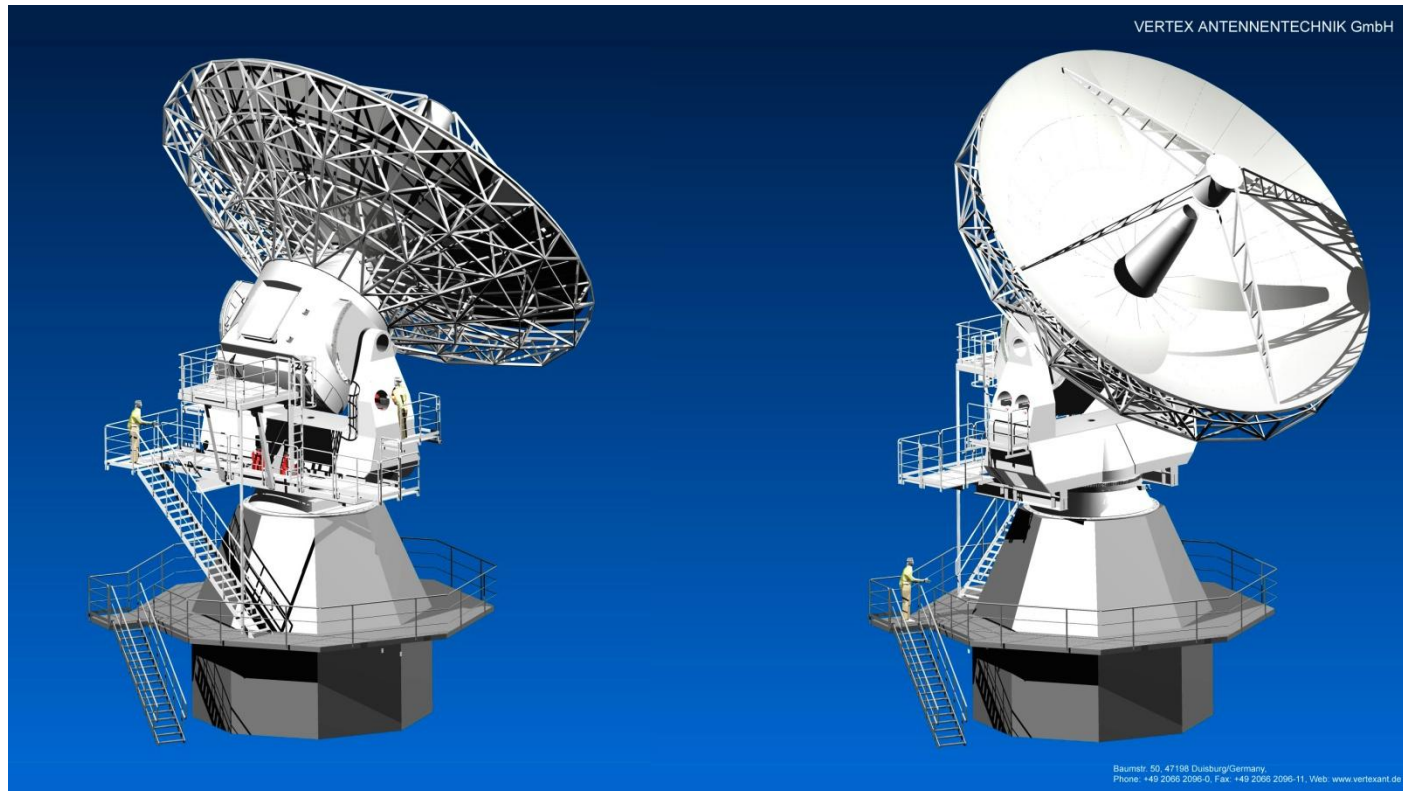




VLBI 2010

Critical Design Points



G. Kronschnabl, A. Neidhardt, K. Pausch, W. Göldi, T. Ekebrand, B. Corey, B. Petrachenko, Jan Kodet, Ch. Plötz;



IVS WG 3 – VLBI2010: Current and future requirements for geodetic VLBI Systems



Goals for a next generation VLBI-System:

- Determination of the relative position better than 1 mm / year
- Continuous observation of the Earth Orientation Parameters
- Very fast generation and distribution of the IVS-Products
 - ➔ continuous, improved UT1 monitoring
 - ➔ Improving of the Celestial Reference Frame (CRF)

Source: IVS WG3 Final Report - <ftp://ivsc.gsfc.nasa.gov/pub/annual-reports/2005/pdf/spcl-vlbi2010.pdf>



- Increasing the numbers of radio sources (up to 1000 scans/day!)
- Improvement of the Delay Observable
- Reduction von systematic errors, i.e. at the electronic devices, the antenna deformation and of the source structures
- Continuous observation rows the whole year (2 antennas)
- Improving of the network geometry
- New improved strategies at the data analysis
- Improved observation schedules
- Additional measurement system, such as a WVR
- Online data transfer via Internet
- Software correlation
- Remote observations



VLBI 2010

What are the requirements for a new observation system to fulfill the VLBI 2010 specifications?

- A fast moving antenna system with an antenna diameter of 12m or more
- A broadband receiving system at least from 2 to 14 GHz, optional a receiver at S-, X-, and Ka-Band
 - ➔ S- und X-Band compatibility (RCP)
 - ➔ stable phase centre and stable reference point
 - ➔ high antenna efficiency and low system temperature
- Improved reference and calibration systems
- New digital data acquisition systems



- **increase the number of observations means:**

- Reducing the observation time (i. e. the Integration time) needs a
 - → better SNR → ($< T_{\text{sys}}$; $>$ higher effective Antenna area)
 - → higher bandwidth (New feeds, new receivers; data boosters)

- Reducing the slewing time needs

- → faster antenna drives
- → more energy consumption
- → a better mechanical construction
- → more attrition and therefore more maintenance

- **reduce systematic errors, such as at the electronic devices, the antenna deformation and additional calibration systems:**

- → better time and frequency reference
- → improving of the phase delay errors, for instance at the cables
- → new improved calibration systems
- → additional measurement systems → Water Vapor Radiometer ?



What are the key points for such an Antenna?

- Antenna diameter size 12m or more
- Fast moving antennas
- Broadband or multiband capability
- Extreme stiff reflector
- High efficiency reflector
- Low path length error
- Very good and stable reference point
- Very stabile towers
- Phase stable cables and cable wraps
- Stable phase centre of the feed (almost frequency independent)
- Remote control
- Energy saving techniques
- Improved time and frequency reference system

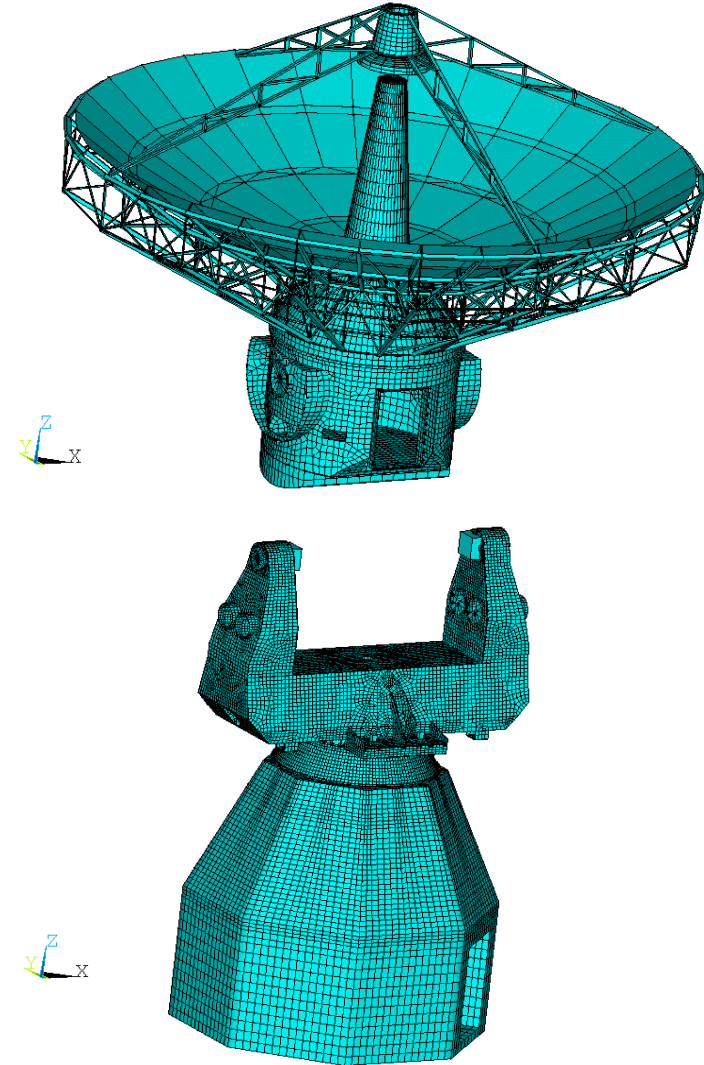
How did we realized that at Wettzell?

Microwave Antenna design:

- Excellent antenna efficiency by the Ringfocal Design a flare angle of about 65
- No blockage by the Subreflector
- Broadband capability > 40 GHz
- Low ground pickup noise
- Mechanical correction of the Subreflector

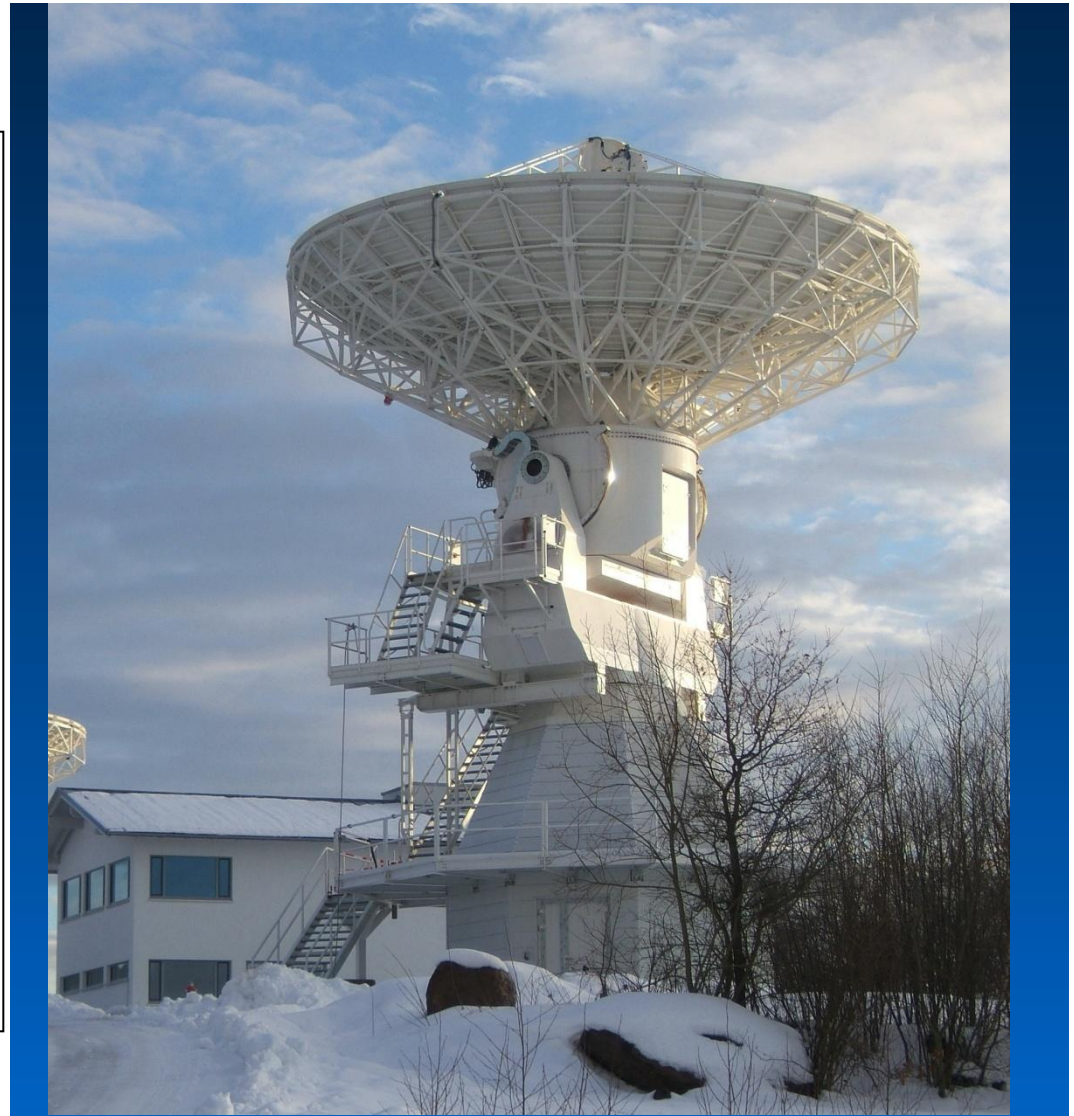
Mechanical Antenne design:

- Very low Path Length Error
- Extreme stiff Main- and Subreflector,
- Extreme stiff Elevation cabin and Azimuth yoke
- Excellent Azimuth- and Elevation bearings
- High resolution hollow shaft encoder
- Very stable towers with a big basement
- Vertical and horizontal axis offset less than 5 arcsec
- Balanced antenna design with counterweights



Technical Data:

- Main reflector: 13.2m
- Ringfocal-Design
- $f/D = 0.29$
- Path Length Error $< 0.3\text{mm}$
- ALMA Mounting with drive velocities of 12 /s in Azimuth and 6 /s in Elevation
- Drive range $\pm 270^\circ$ & 115°
- Balanced antenna design
- Excellent bearings
- 27Bit Encoder : 0.0003 resolution
- Subreflector adjustable by a Hexapod



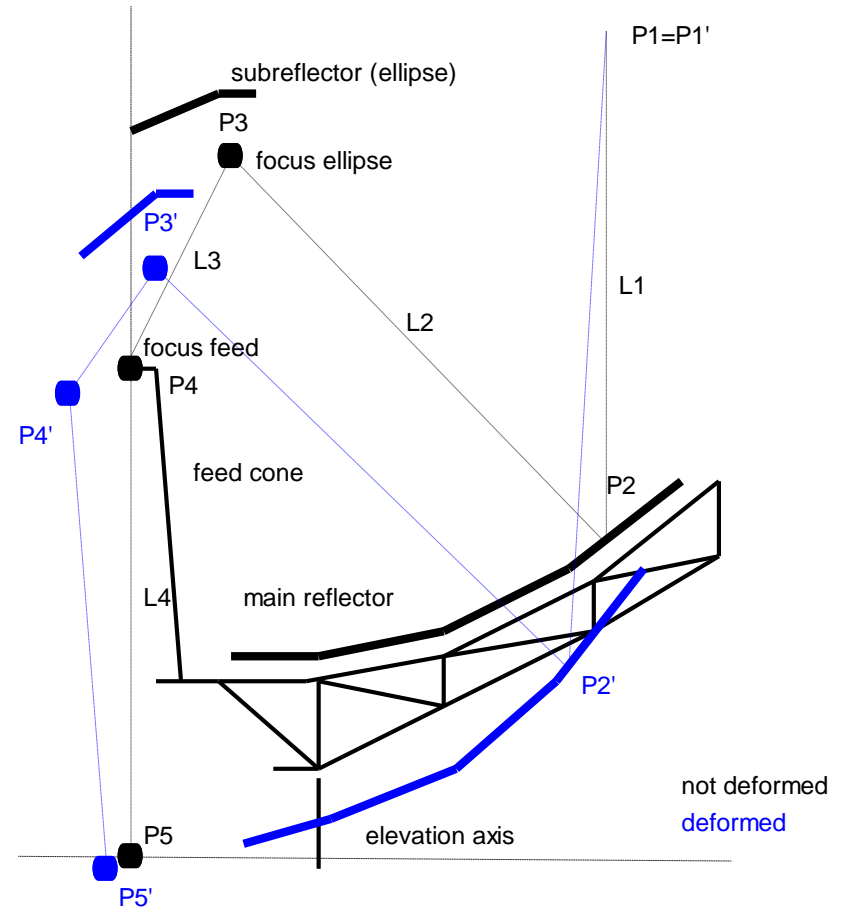
The TTW-Antenna is designed for a Path Length Error of less than 0.3mm !!

- L1 = distance main axis – reflector surface
- L2 = distance reflector surface – subreflector
- L3 = distance subreflektor – feed focus
- L4 = distance feed focus – axis intersection point

Definition: Path Length Error

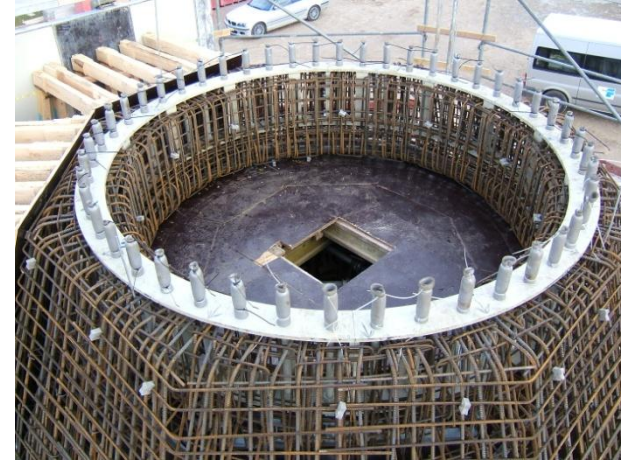
- $L_{not_deformed} = L1 + L2 + L3 + L4$
- $L_{deformed} = (L1 + dL1) + (L2 + dL2) + (L3 + dL3) + (L4 + dL4)$
- $L_Error = L_{deformed} - L_{not_deformed}$

$$PathLengthError = \frac{\sum_{i=1}^{192} A_i \cdot PathLengthError_i}{\sum_{i=1}^{192} A_i}$$



Source: Vertex Design Review; Dez. 2008

- Basement up to 6m in the ground
- Thick concret walls containing tons of steel



Zul. $\varphi_{OKT} \leq 0,0005^\circ$

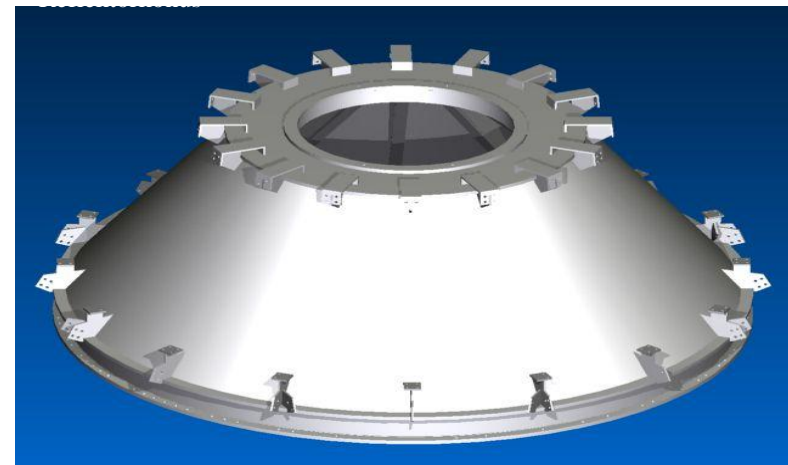
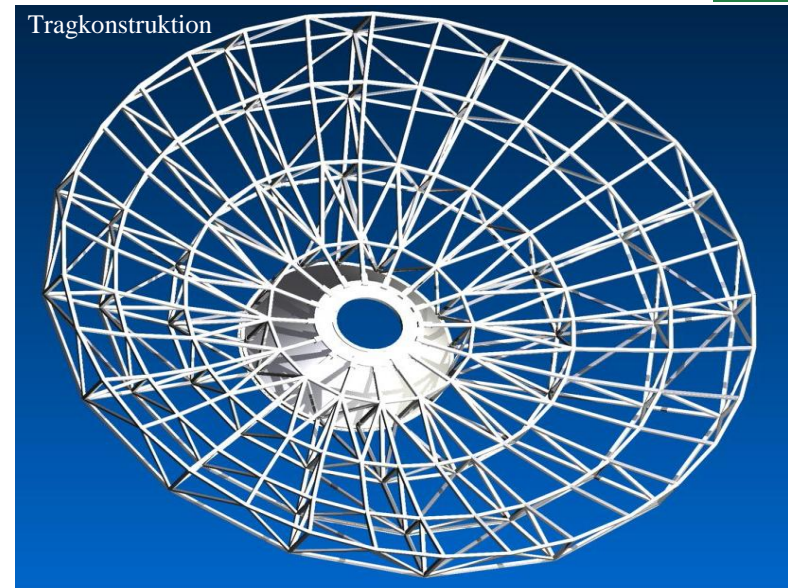
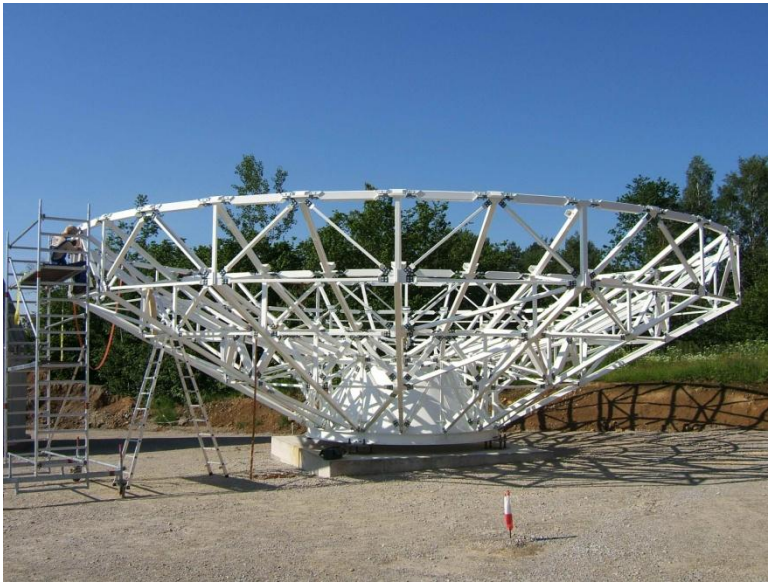
Unter Wind ≤ 40 km/h, Böen bis 50 km/h

Zul. $\varphi_{OKT} \leq 0,0015^\circ$

Unter Temperatureinwirkung



TWIN – Radiotelescope: Main reflector construction



Bezeichnung	FE-Analyse		Fotogrammetrie	
	0° EI	90° EI	0° EI	90° EI
Oberflächenfehler RMS [μm]	149	131	128	123

Tab. 2-13: Oberflächenfehler des Hauptreflektors

Bezeichnung	Transformierte FE-Daten	Fotogrammetrie
Verschiebung y_{Hr} [mm]	0.71	0.51
Verkipfung $\phi_{x,\text{Hr}}$ [Grad]	-0.021	-0.019

Tab. 2-14: Vergleich von Hauptreflektordaten bei der Elevationsänderung 90°->0°

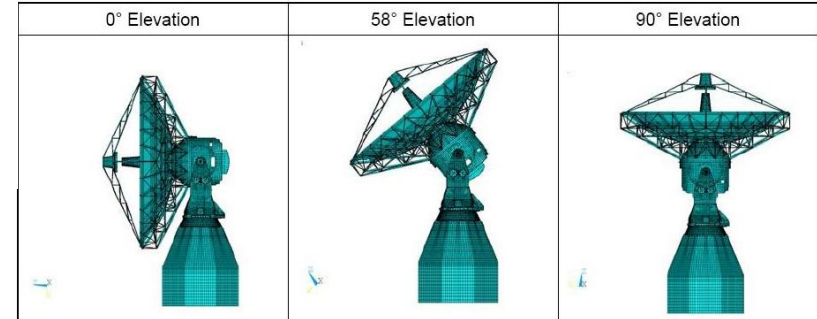
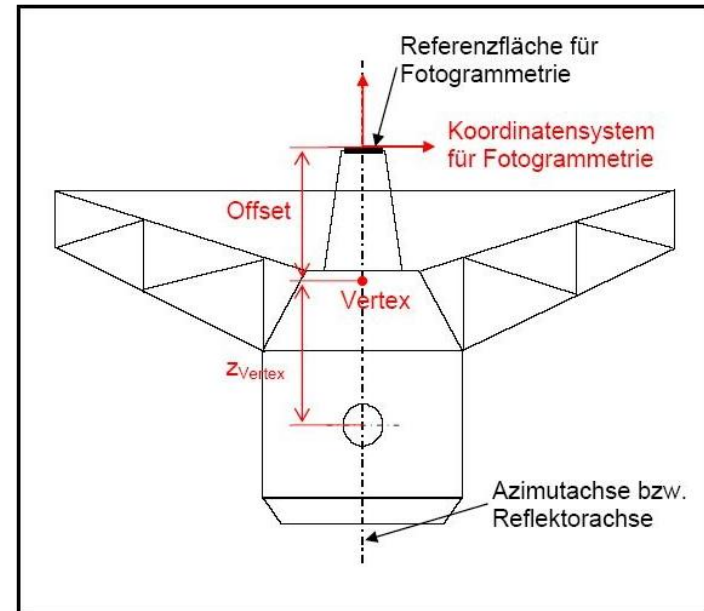
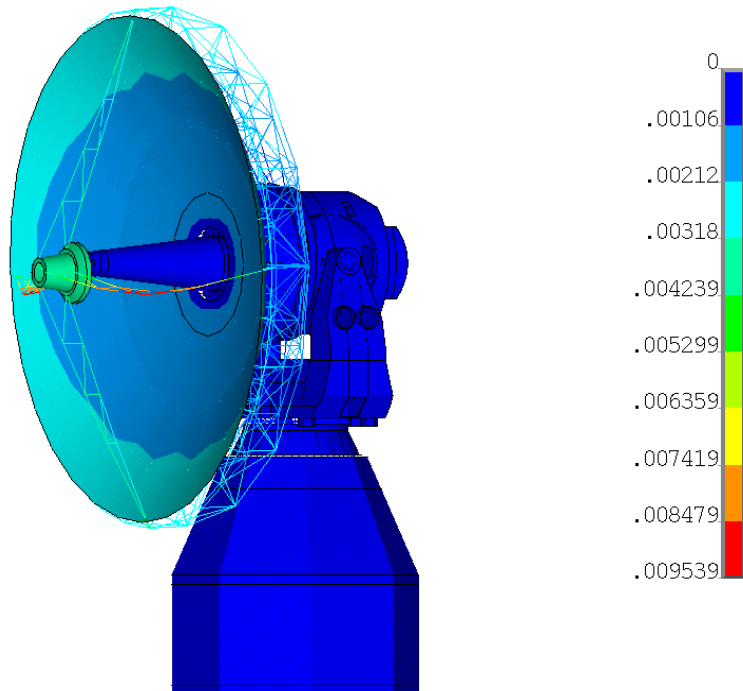


Abb. 2-1: FE-Modelle der TTW-Antenne bei 0 Grad, 58 Grad und 90 Grad Elevation



Prinzipalskizze zur Referenzfläche für die Fotogrammetrie

Elevationsänderung:	90° -> 58°	Wert	Einheit
Verschiebung in x-Richtung		-0.02	[mm]
Verschiebung in y-Richtung		1.24	[mm]
Verschiebung in z-Richtung		0.26	[mm]
Verkippung um die x-Achse		0.003	[deg]
Verkippung um die y-Achse		0.002	[deg]

Tab. 10-1: Messprotokoll der gemessenen Subreflektorverschiebungen und -verkippungen bei Elevationsänderung von 90° nach 58°

Elevationsänderung:	90° -> 0°	Wert	Einheit
Verschiebung in x-Richtung		-0.07	[mm]
Verschiebung in y-Richtung		2.30	[mm]
Verschiebung in z-Richtung		2.20	[mm]
Verkippung um die x-Achse		0.006	[deg]
Verkippung um die y-Achse		0.005	[deg]

Tab. 10-2: Messprotokoll der gemessenen Subreflektorverschiebungen und -verkippungen bei Elevationsänderung von 90° nach 0°

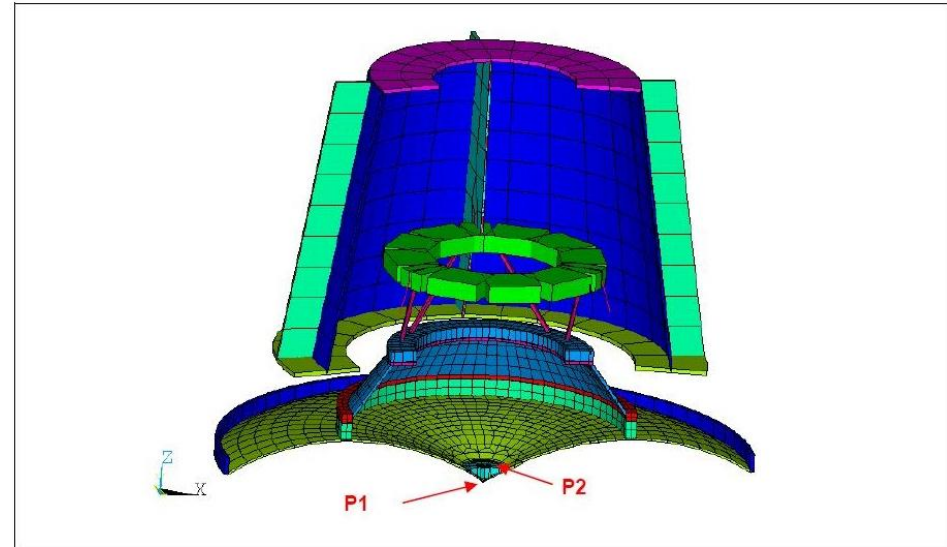


Abb. 2-3: Kopfteil mit Subreflektor und Auswertepunkte P1 und P2

Bezeichnung	Transformierte FE-Daten	Messung Mittelwert
Verschiebung y_{Sr} [mm]	2.64	2.30
Verschiebung z_{Sr} [mm]	1.73	2.07
Verkippung $\varphi_{x,Sr}$ [Grad]	0.005	0.007

Tab. 2-15: Vergleich von Subreflektordaten bei der Elevationsänderung 90°->0°

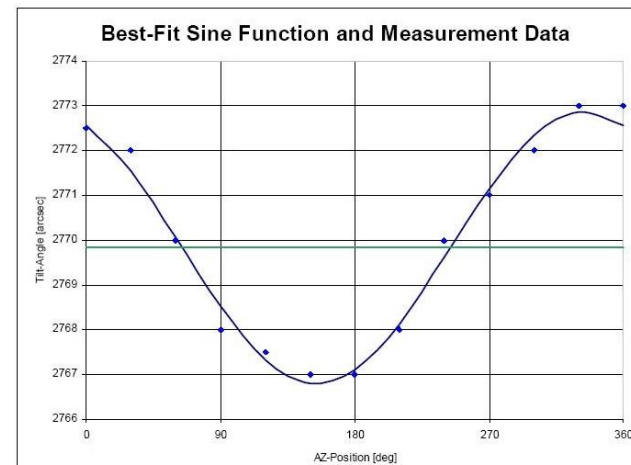
- LF1 = Loading at 58 without Subreflectorcontroller
- LF2 = Loading at 58 with Subreflectorcontoller
- LF3 = Load with wind speed 40km/h ahead
- LF4 = Load with wind from one side

LF	Gain 33 GHz [dBi]	Δ -Gain [dB]	η	RMS σ [mm]
nominell	72.412			
1	72.193	-0.219	0.951	0.16
2	72.318	-0.094	0.979	0.11
3	72.408	-0.004	0.999	0.02
4	72.409	-0.003	0.999	0.02

Tab. 3-1: Betrachtung des Antennengewinns



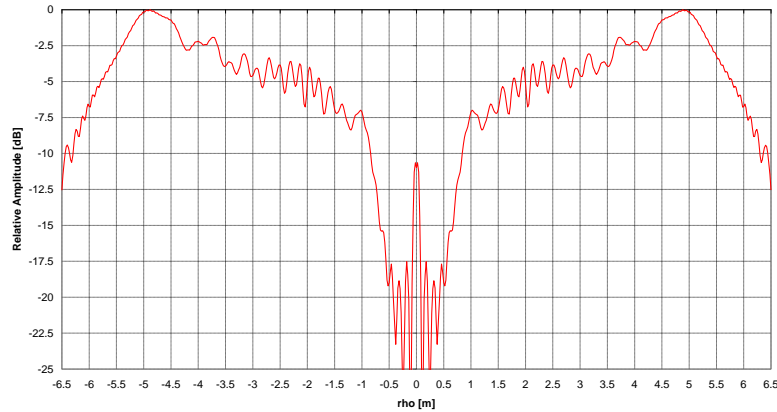
Measurements of the vertical and horizontal axis and the intersection point



Darstellung von Messwerten der Azimutachsmessung sowie deren Bestfit-Sinusfunktion

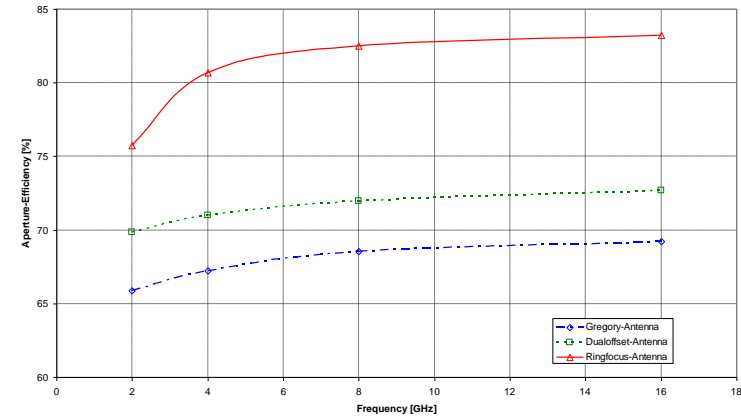
Distribution of the radiated energy

13.2m Ring Focus Antenna
Aperture Field Distribution, $f = 5$ GHz



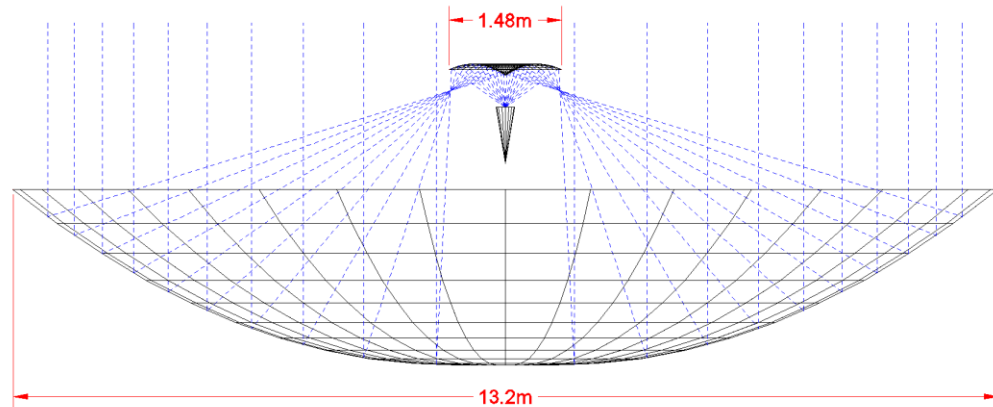
Effective beam efficiency

13.2m Antennas with Gaussian Beam Feeds (-12dB at Subreflector Rim)
Aperture Efficiency



Ringfocal-Design

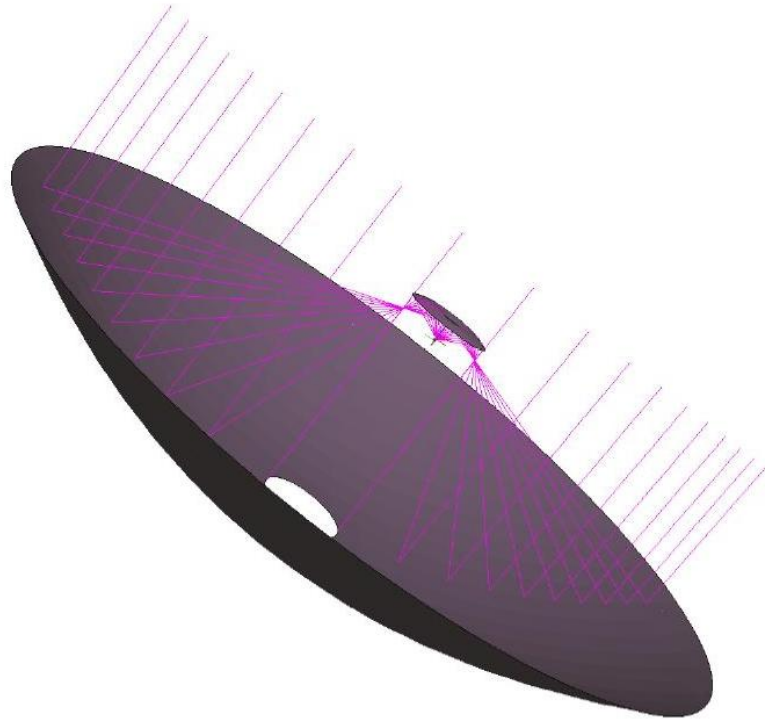
- Dual-Reflector receiving system
- optimal for large flare angles
- no blockage by the subreflector
- high illumination efficiency
- the feed horn is prevented by radiation from the sun



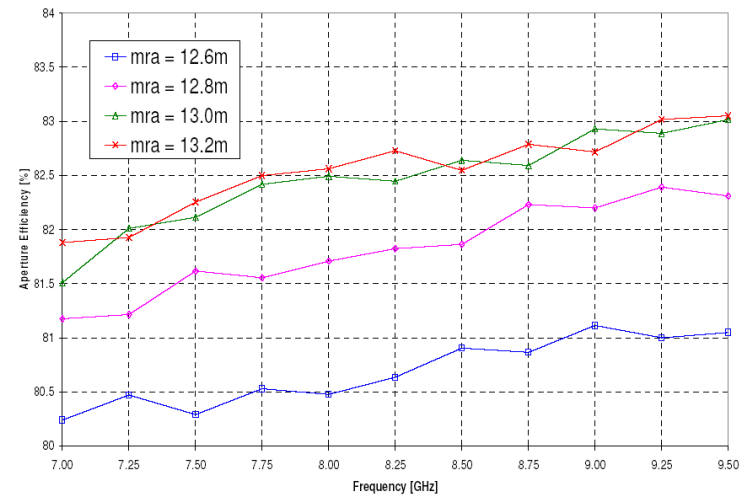
Source: Willi Göldi, Mirad; FRFF-Workshop 2009, Wettzell



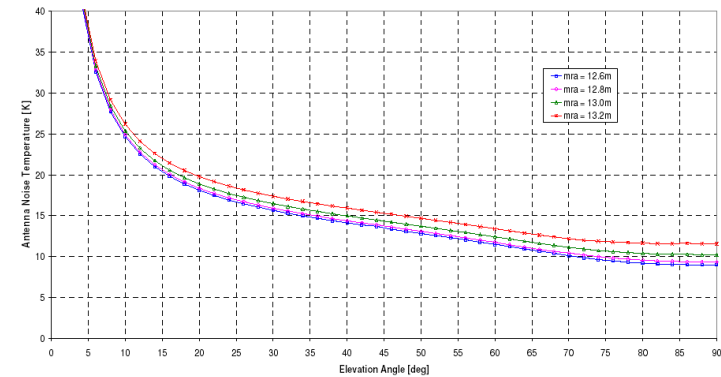
Antenna efficiency and ground noise pickup



TTW Aperture Efficiency with Different Main Reflector Illumination
with Gaussian Feed, X-Band



with Gaussian Feed, X-Band, $f = 8.25$ GHz

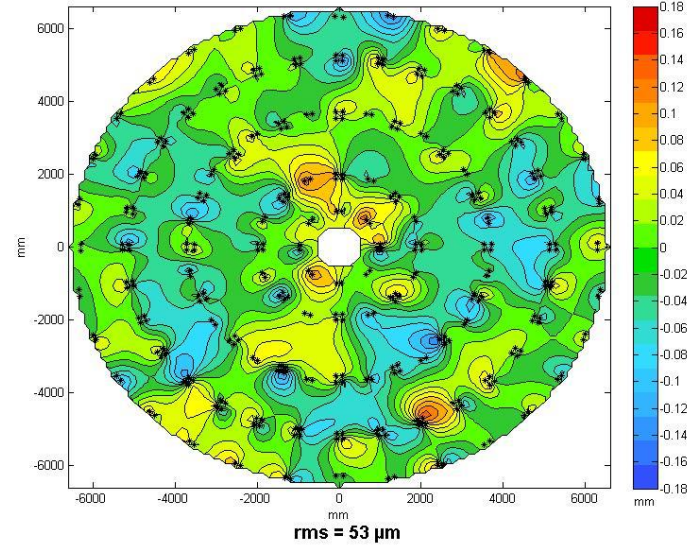




Main reflector



Antenne #1: Abschlussmessung (58° Elevation)



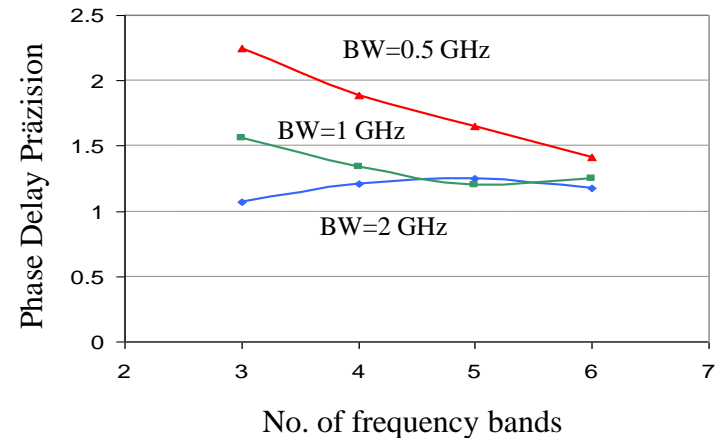
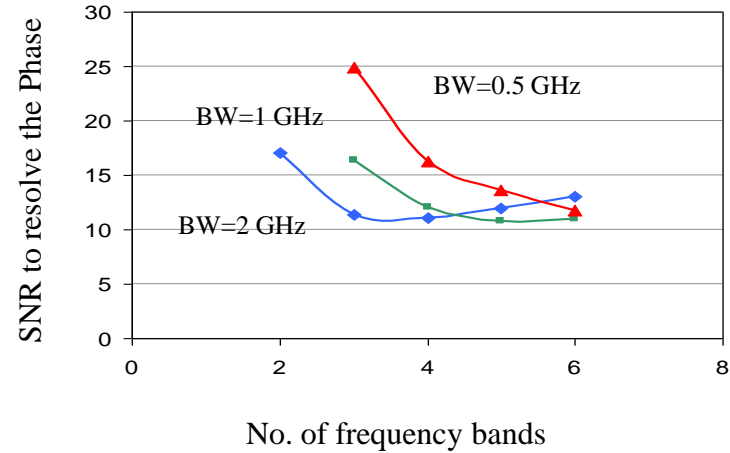
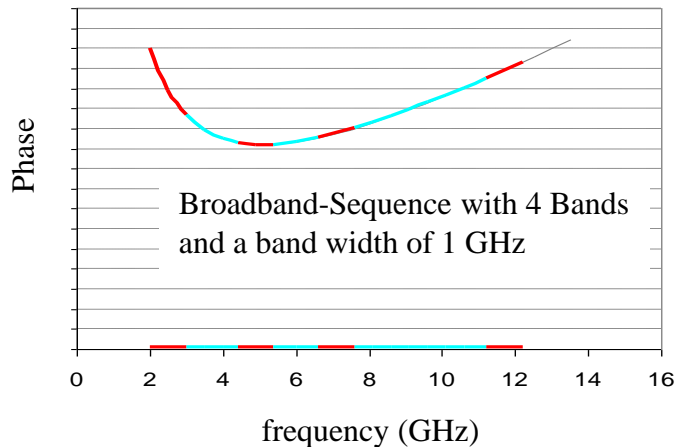
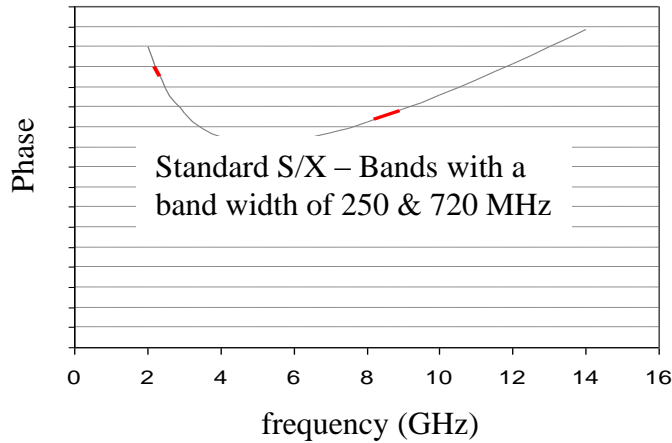
FAT Subreflector





VLBI 2010: Broadband Delay

- Delay precision target: 4-ps (in reality larger)
- Frequency range: 2-18 GHz; (in reality 14 GHz with reasonable efficiency)
- Number of bands: 4
- Bandwidth per band: 1 GHz



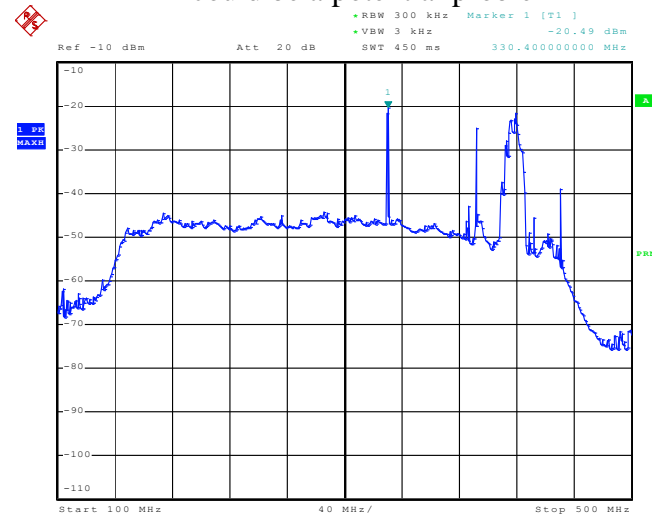
Source: B. Petrachenko: Broadband Delay Tutorial, FRFF Wettzell 2009

Wideband Disadvantages?

Subreflector is very close to the feed



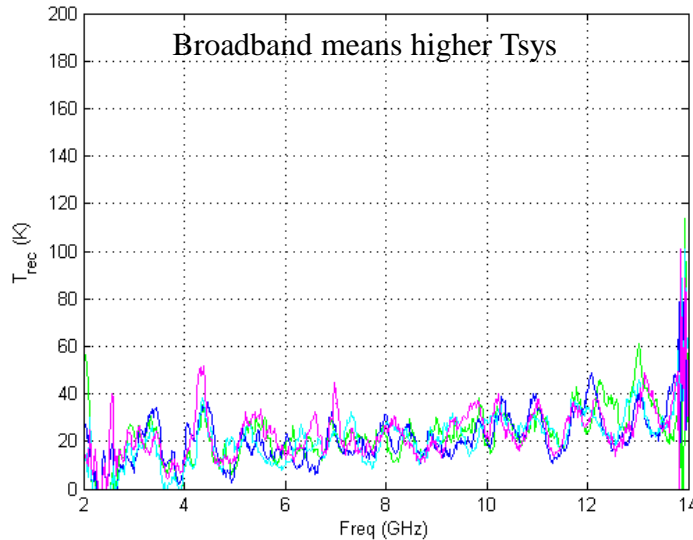
RFI could be a potential problem



Date: 10.MAR.2009 13:19:42

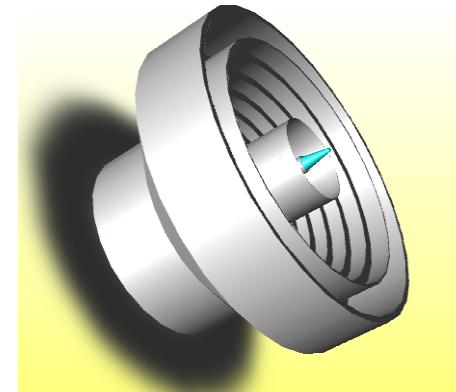
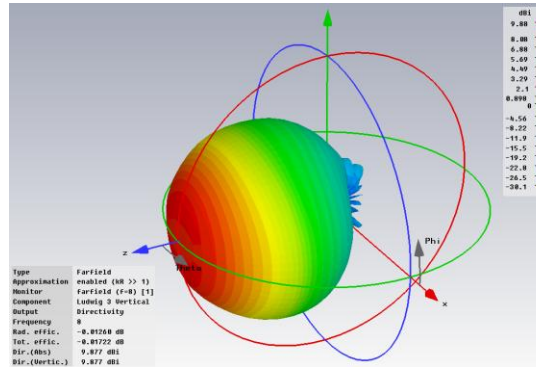
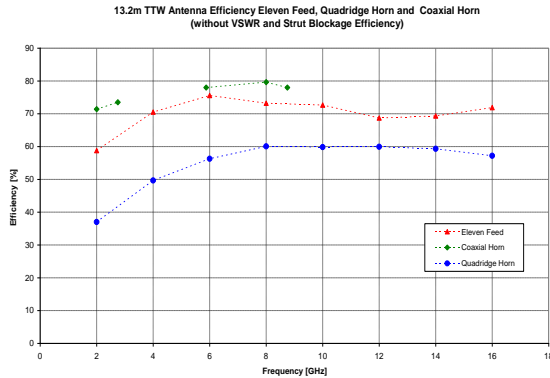
- Huge amount of data
- Data handling and storage
- Correlation time increases

Broadband means higher T_{sys}

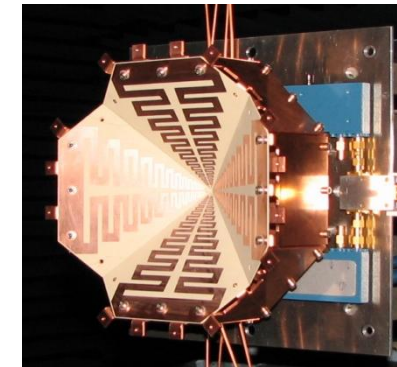
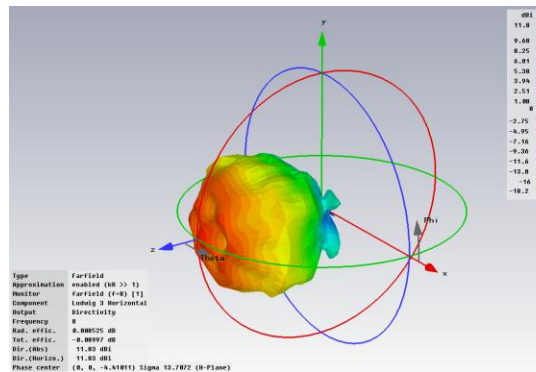
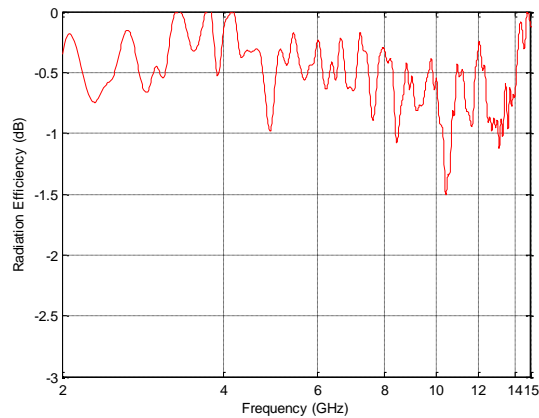


A broadband feed has a wide flare angle! (65°)

Triband Coaxial-Feed



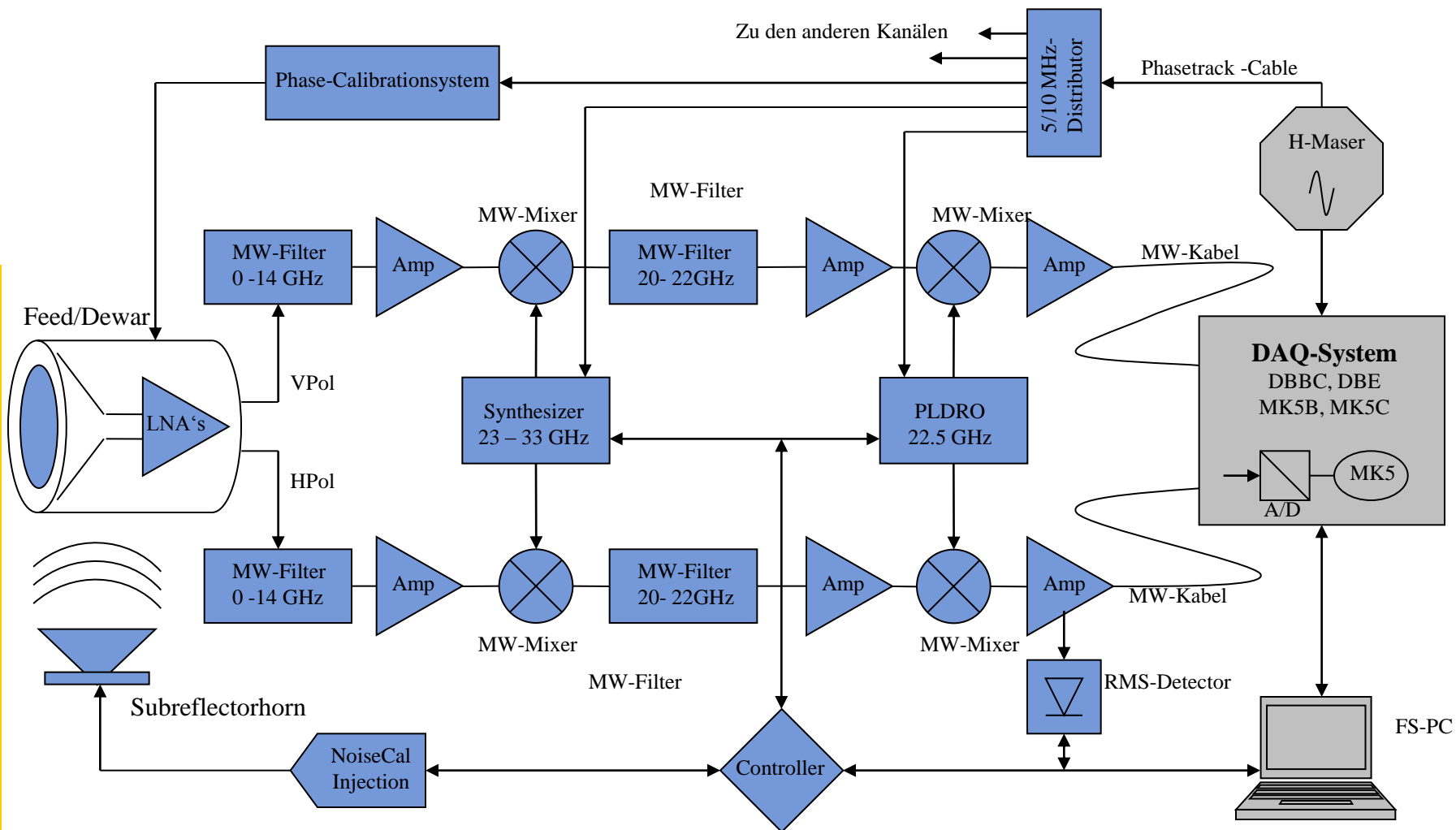
Eleven Feed



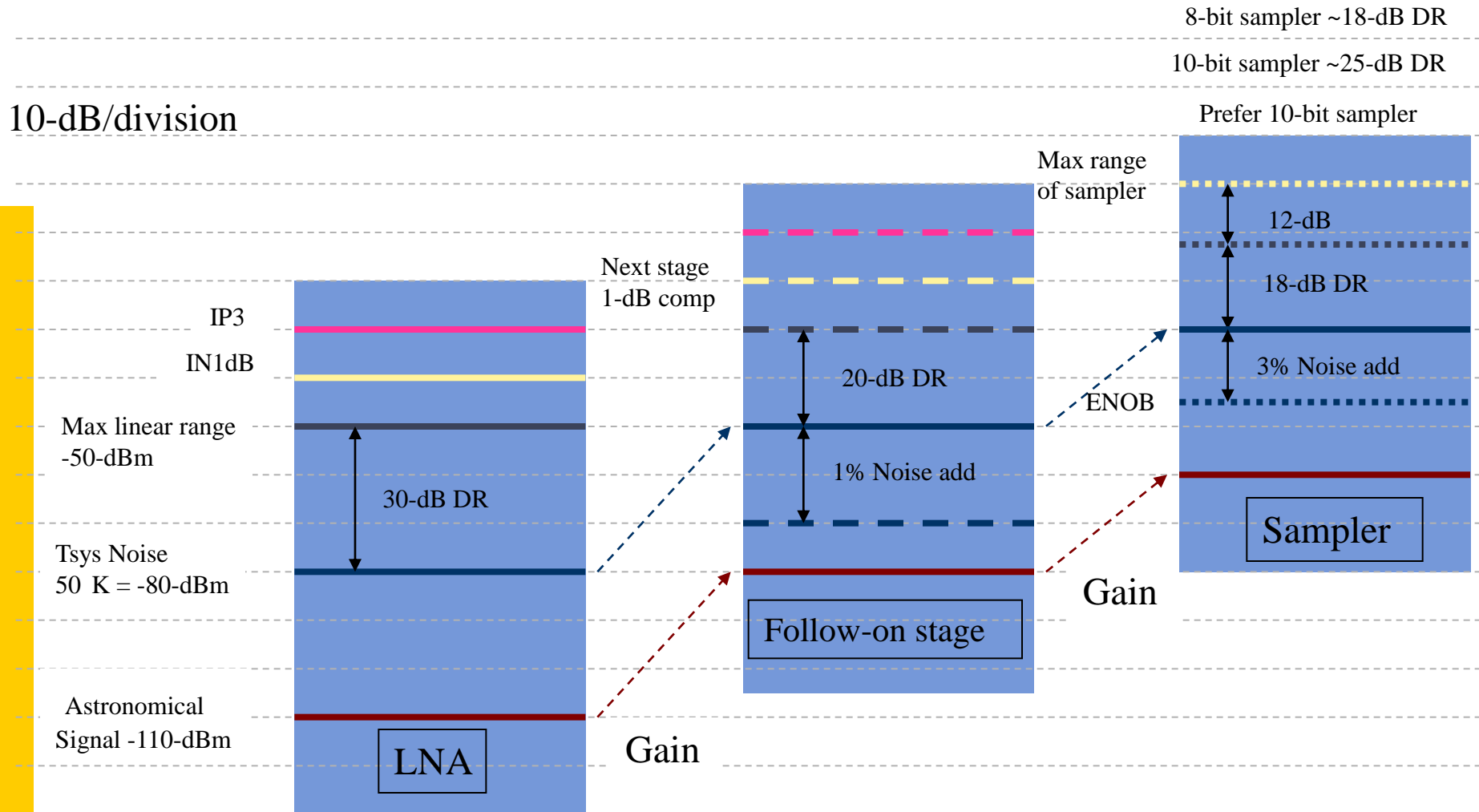
Source: Willi Göldi, Mirad; S. Kildal, Chalmers Univ.; FRFF-Workshop 2009, Wettzell



Receiving-System: Wideband-Receiver



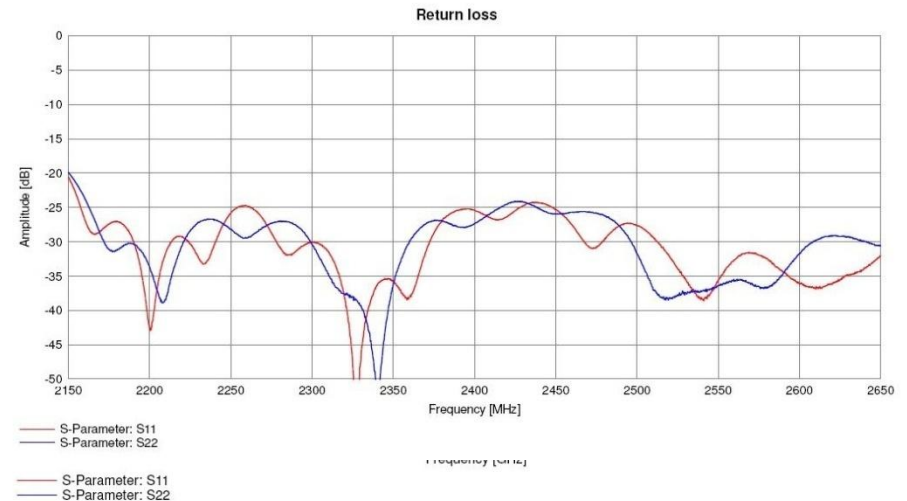
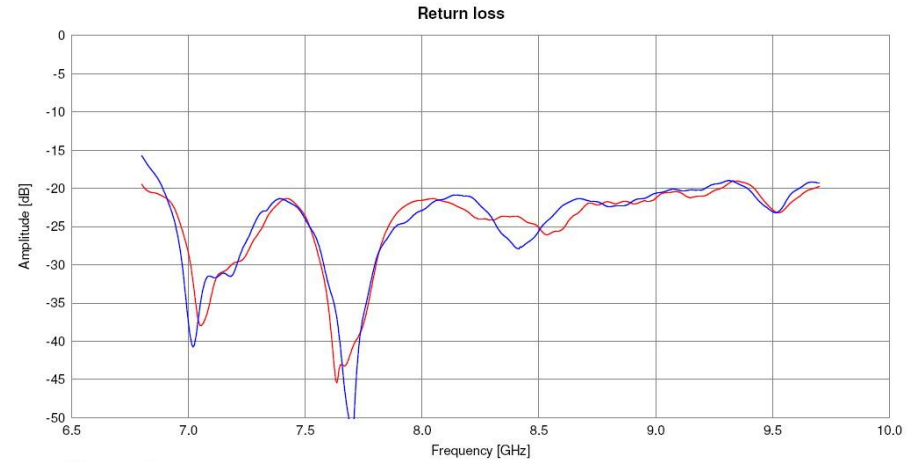
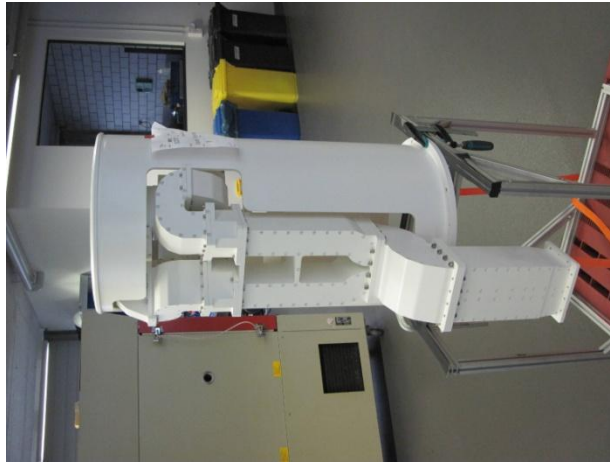
Receiver Dynamic Range



Source: Brian Corey; Tecspec-Meeting 2012



First measurement results of the new Triband-Feed

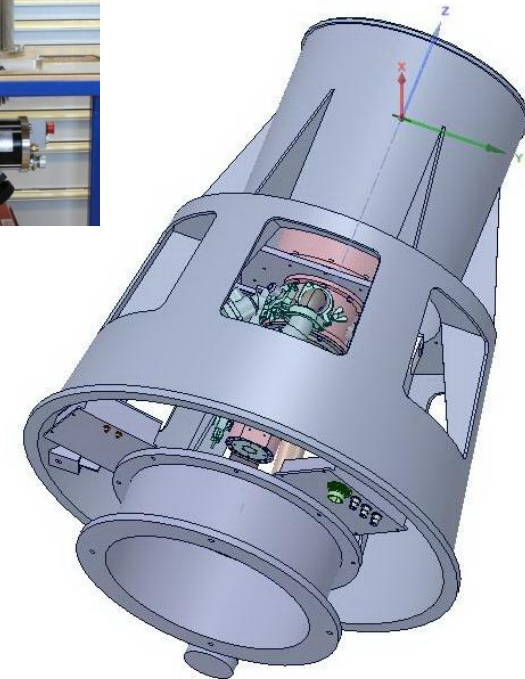


Design Status Elevenfeed

Elevenfeed with Cyrosystem



Elevenfeed Dewar



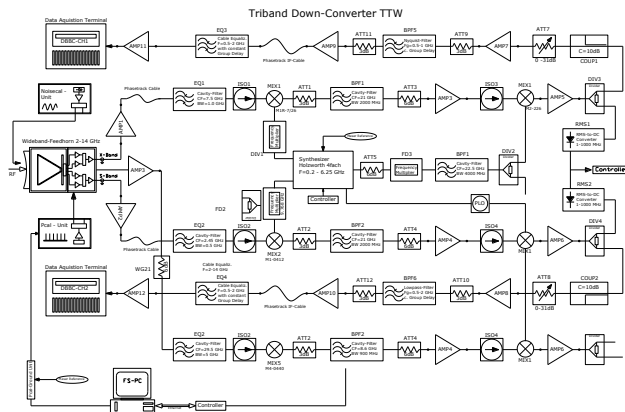
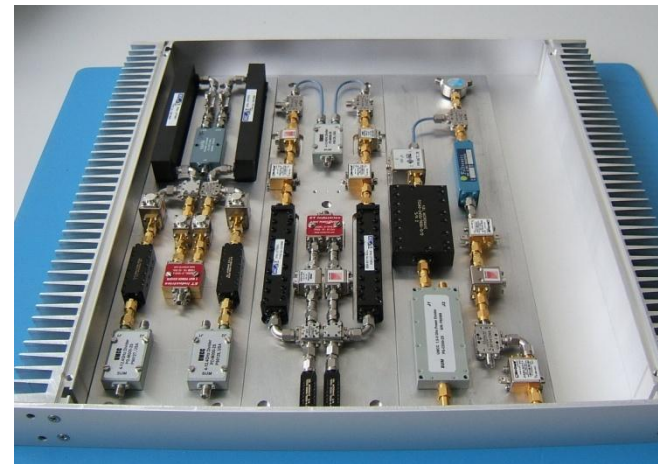
Receiver Cone

TTW1 Receiver Design

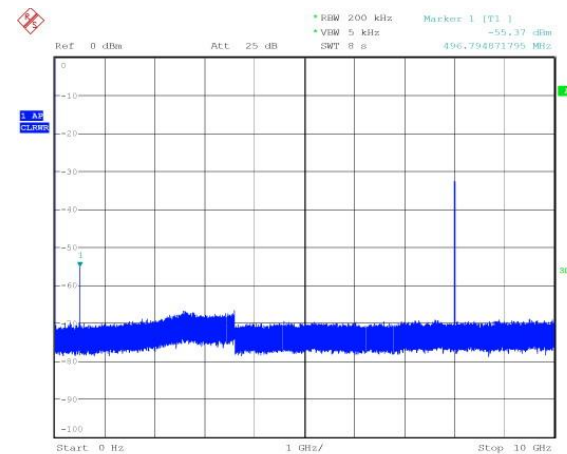
S/X-Receiver



S/X/Ka-Band Receiver



Tri-Band Receiver



S-Band Up/Down-Converter Measurement



Absolute Time in VLBI

VLBI is *the* space-geodetic technique for measuring UT1 = Earth rotation angle relative to universe.
Time error impacts UT1 directly and should be: $< 0.1 * 500\text{-ns} = \sim 50\text{-ns}$.

This includes:

- Time labelling of sampled data $< \sim 10\text{-ns}$
- Delay of the PCAL uplink cable $< \sim 5\text{-ns}$
- Delay of the cable for sampling clock epoch $< \sim 5\text{-ns}$
- Delay of the GPS cable. $< \sim 5\text{-ns}$
- Absolute GPS time $< \sim 50\text{-ns}$

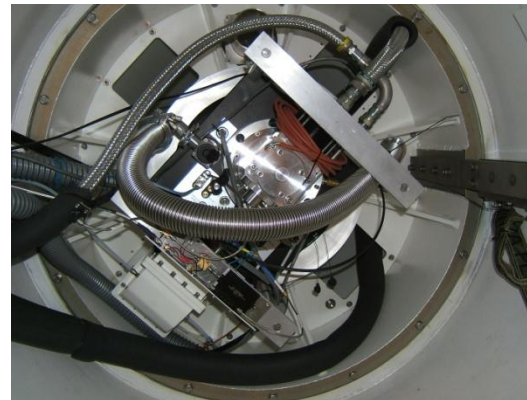
The UT1 **precision** in VLBI2010 expected to be $\sim 0.5 \mu\text{s}$.

To achieve $0.5 \mu\text{s}$ UT1 accuracy, we need to know station time to well under $1 \mu\text{s}$ relative to UTC.

A value of 50 ns would be good enough, but better accuracy is possible? (source: B. Corey Tecspec-meeting)



Station time reference point !



100MHz →

← 1PPS



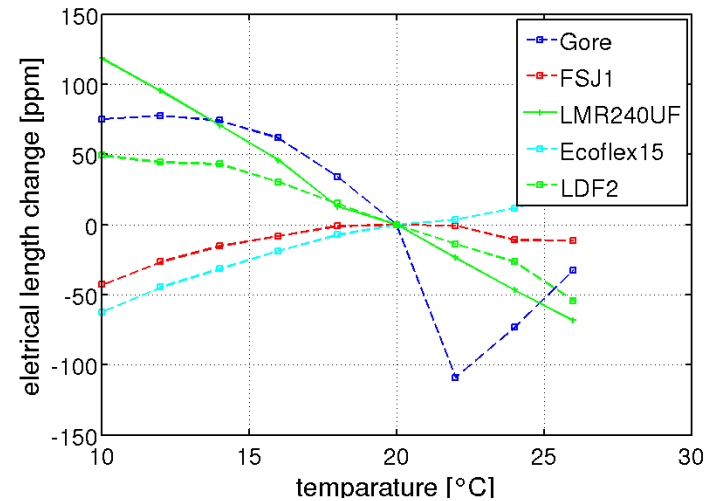
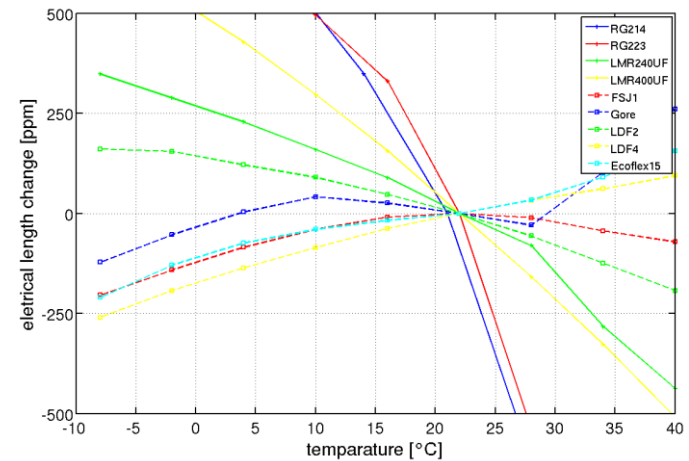
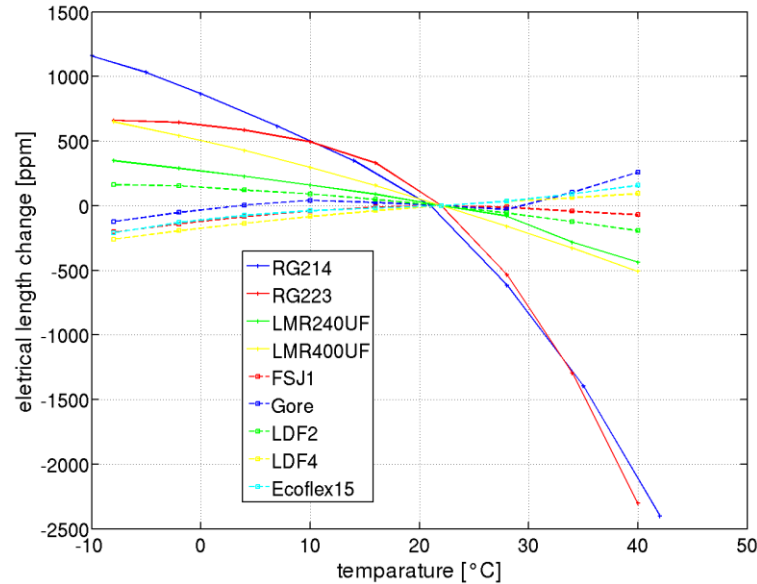


- Cable carrying pcal reference from maser to frontend must be stable in absence of cable cal.
- Specs on pcal reference cable stability in absence of cable measurement system:
 - < 0.3 ps variations that are dependent on antenna orientation
 - Allan std dev < 10^{-15} @ 50 minutes
 - On other time scales, ASD scales with typical maser performance.
- ASD spec also applies to any buffer amp that drives the pcal reference signal.
 - Be sure buffer amp is not sitting in front of air conditioning vent!
- If pcal is absent, effect of variations in sampler clock delay is scaled down by frequency ratio at sampler input and at RF.
- General good practice and possibility of pcal failure argue for using cable with low temperature coefficient for all long cable runs carrying frequency reference.
- Orientation-dependent length variations of pcal ref cable must be < 0.3 ps in absence of cable measurement system.

<0.5 ps	LMR-240 coax	1 turn with 10-cm radius
<0.5 ps	LMR-400 Ultraflex coax	1 turn with 8-cm radius
<0.5 ps	Single-mode fiber in “loose bundle”	1 turn with 8-cm radius

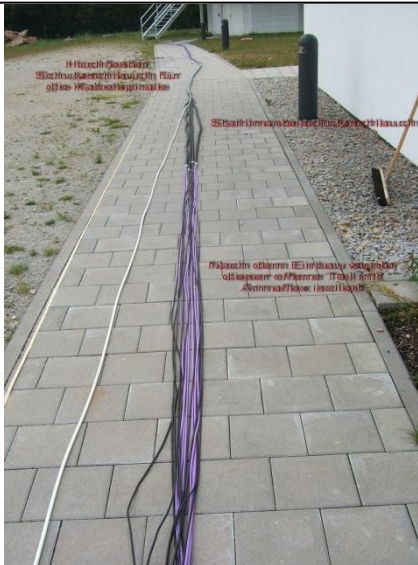
Source: Brian Corey; Tecspec-Meeting 2012

Tests of various cables in the temperature chamber



Minimizing temperature-driven cable length variations

- Use cable with low temperature coefficient.
 - Example:
 - 80-m LMR-400 @ 5 MHz: $80\text{m} \times 4\text{ns/m} \times 3\text{ppm/K} = 1.0 \text{ ps/K}$ (upper limit) (AEER)
 - 20-m LMR-400UF @ 5 MHz (cable wrap): $20\text{m} \times 4\text{ns/m} \times 9\text{ppm/K} = 0.7 \text{ ps/K}$ (AEER)
 - 75-m FSJ1 cable @ 100 MHz : $75\text{m} \times 4\text{ns/m} \times 5\text{ppm/K} = 1.5 \text{ ps/K}$ (J. K)
 - 32-m LMR-240UF @ 100 MHz (cable wrap): $\sim 1.5 \text{ ps/K}$ (J. Kodet)
- Add thermal insulation around exposed cables to increase thermal time constant.
 - Thermal time constant of bare LMR-400 is ~ 30 minutes.
- Bury cables from control room to antenna pedestal.





- Primary function: Measure time variations of instrumental phase vs. frequency.
- Secondary functions:
 - Infer T_{sys} variations from phase cal amplitude.
 - Phase/gain equalization for circular polarization generation from linear pol.
- Phase differences between channels will be far more stable in VLBI2010 than in S/X VLBI, thanks to digital IF-to-baseband conversion in FPGAs.
- But phase cal is still needed in VLBI2010 to measure
 - LO phase drifts between bands
 - Phase/delay drifts in RF/IF analog electronics As RF bandwidth increases, pulse intensifies.

Problem:

- With insufficient analog headroom, pulse drives electronics into nonlinear operation. → spurious signals generated that corrupt undistorted pcal signal
- Options to avoid driving electronics into saturation:
 - Reduce pulse strength
 - Phase cal SNR reduced → noisier phase extraction
 - More prone to contamination by spurious signals
 - Reduce pulse strength *and* increase pulse repetition rate to 5 or 10 MHz
 - Fewer tones spaced 5 or 10 MHz apart
- With 5 or 10 MHz rep rate, baseband tone frequencies can differ from channel to channel when channel separation = 2^N MHz.
 - Fringe-fitting is more complicated if only one tone per channel is extracted .
 - Software solution: Use multiple tones per channel and correct for delay within each channel, as well as between channels.

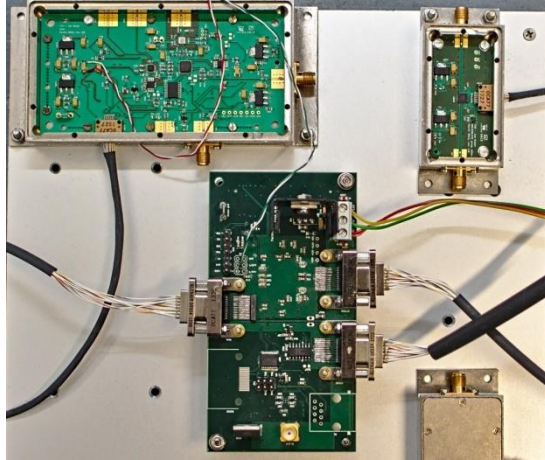
Source: Brian Corey; Tecspec-Meeting 2012



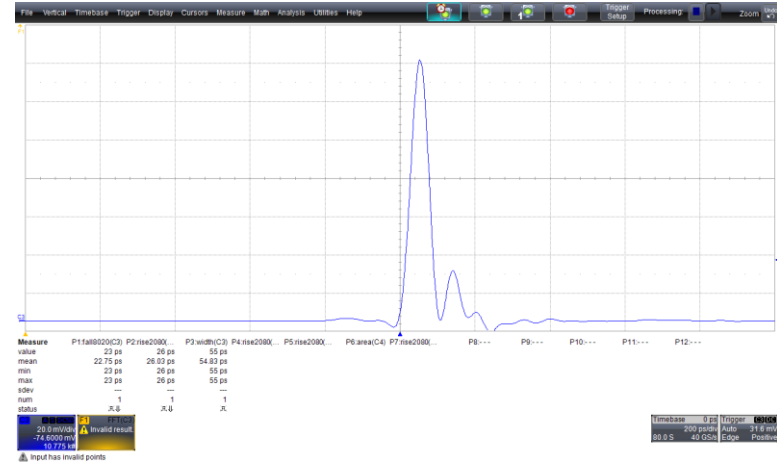
- As RF bandwidth increases, pulse intensifies.
- With insufficient analog headroom, pulse drives electronics into nonlinear operation. → spurious signals generated that corrupt undistorted pcal signal
- Options to avoid driving electronics into saturation:
 - Reduce pulse strength
 - Phase cal SNR reduced → noisier phase extraction
 - More prone to contamination by spurious signals
 - Reduce pulse strength *and* increase pulse repetition rate to 5 or 10 MHz
 - Fewer tones spaced 5 or 10 MHz apart
- With 5 or 10 MHz rep rate, baseband tone frequencies can differ from channel to channel when channel separation = 2^N MHz.
 - Fringe-fitting is more complicated if only one tone per channel is extracted .
 - Software solution: Use multiple tones per channel and correct for delay within each channel, as well as between channels.
- General recommendation: peak pcal pulse power / P1dB < -10 dB

Source: Brian Corey; Tecspec-Meeting 2012

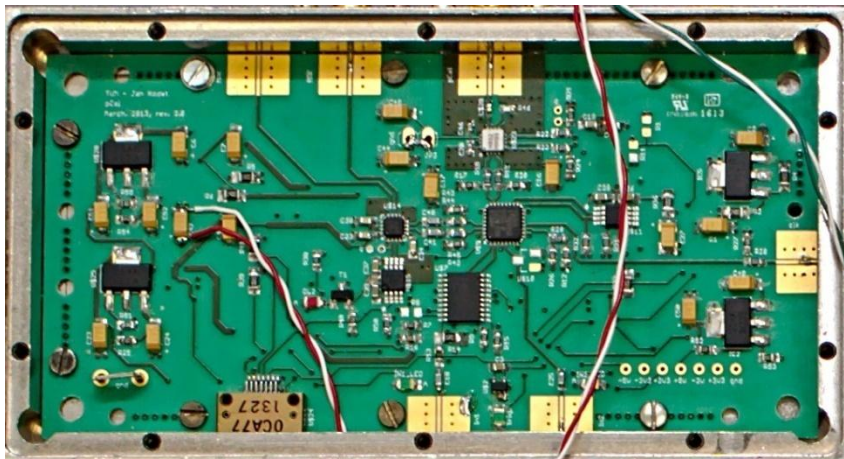
Pcal Development-System



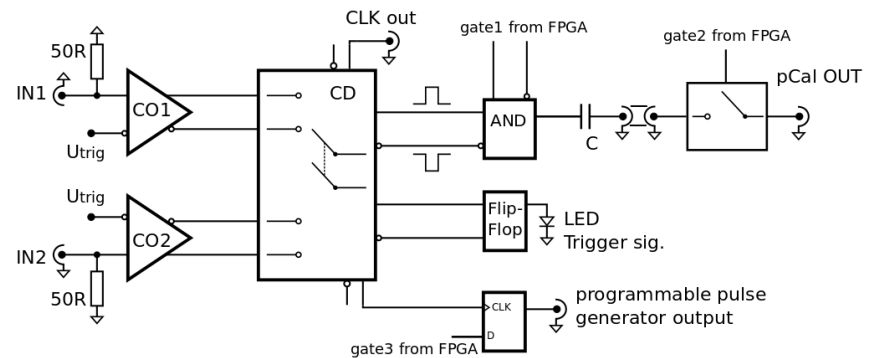
Phasecal-Pulse in Time Domain



Phasecal-PCB

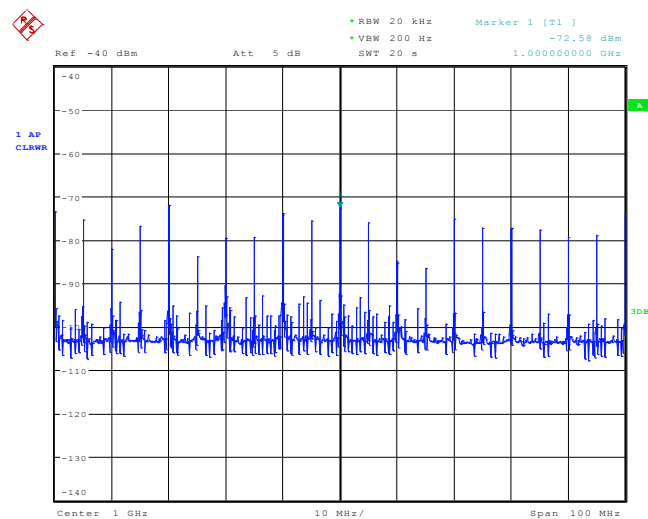
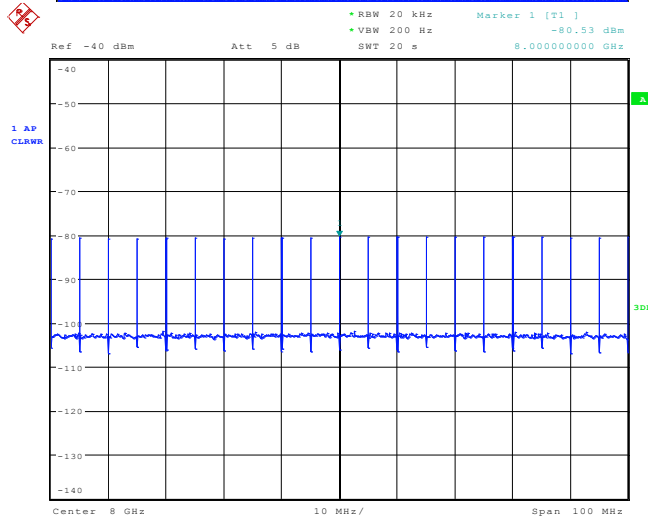
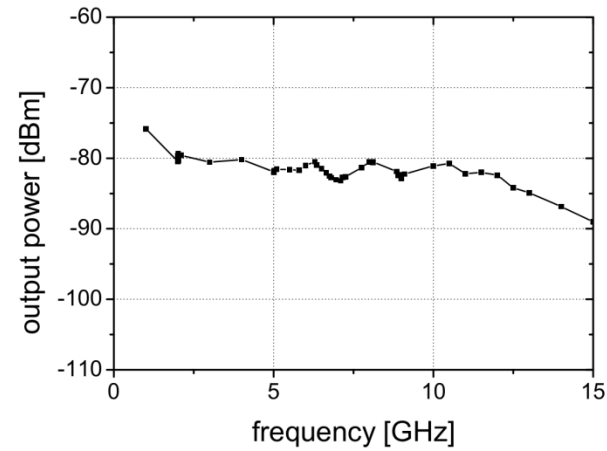
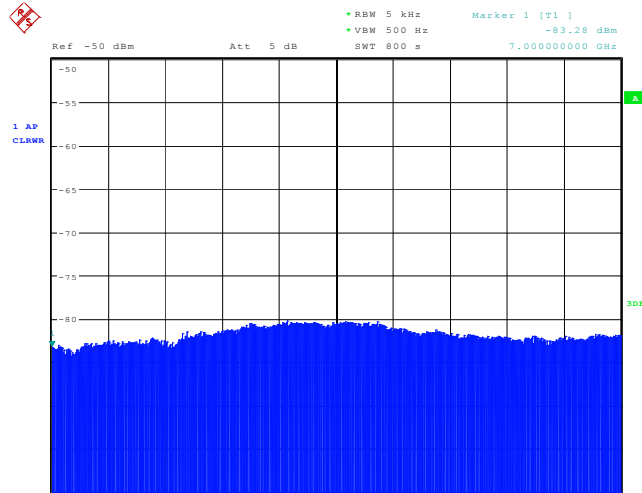


Blockdiagram New Phasecal Unit





New TTW Phasecal Power Level

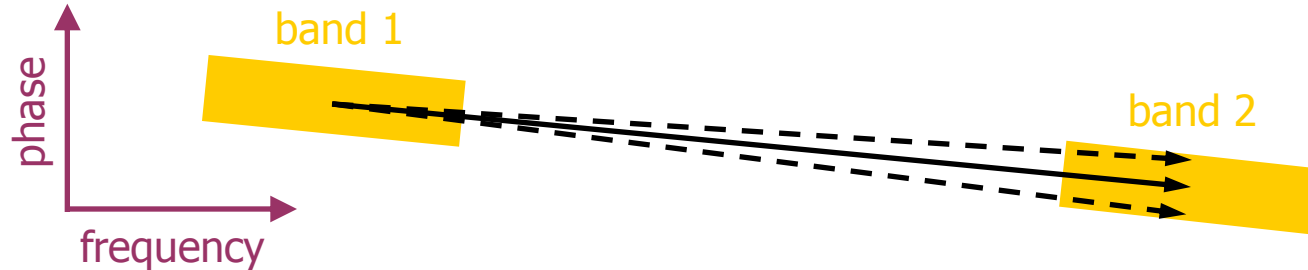


Date: 16.JAN.2014 17:53:02

Date: 16.JAN.2014 18:18:00



Spec for spurious signals independent of antenna orientation



- Case 1: To create bbdelay, must extrapolate phase between two bands up to 5 GHz apart.
 - Require extrapolated phase to be precise to $< 1/10$ radian.
 - \rightarrow delay error $< 0.1 / (2\pi \times 5 \text{ GHz}) = 3 \text{ ps}$
 - 3-ps delay = 0.02 radian (1) over 1 GHz, or 0.01 radian over 500 MHz
- Case 2: One fall-back option is to use group delay over 3 contiguous bands.
 - For SNR = 20, $\sigma(\text{group delay}) \approx 10 \text{ ps}$.
 - Want instrumental error $\ll \sigma$. \rightarrow instrumental error $< 1 \text{ ps}$.
 - 1-ps delay = 0.02 radian over 3 GHz
- Specification for spurs that do not depend on antenna orientation:
 - Sufficient condition: spurs $< -40 \text{ dB}$ relative to pcal
 - Necessary condition: delay error $< 3 \text{ ps}$ over 1 GHz and $< 1 \text{ ps}$ over 3 GHz



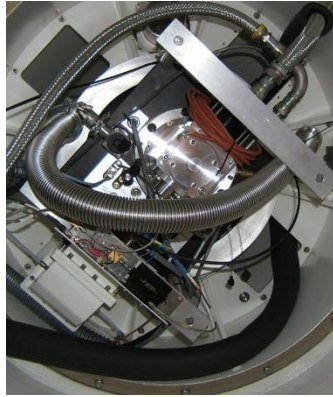
Spec for spurious signals dependent on antenna orientation

- Spurs that vary systematically with antenna orientation need their own spec.
- Possible origins:
 - Varying multipath affecting pcal radiated (intentionally or not!) into feed
 - Elevation-driven thermal variations in pulse generator
- VLBI2010 goal is 1-mm 3-D station position accuracy in 24 hours.
- Orientation-dependent systematic errors...
 - map into station position, and therefore
 - should be kept $< 0.1 \text{ mm} = 0.3 \text{ ps}$.
- 0.3 ps error can arise from spur-induced phase error of
 - 0.004 radian at 2 GHz (broadband delay case), or
 - 0.006 radian change over 3 GHz (case 2 of previous slide).
- Specification for spurs that vary with antenna orientation:
 - Sufficient: spurs $< -50 \text{ dB}$ relative to pcal
 - Necessary: phase error $< 0.004 \text{ radian}$ & delay error $< 0.3 \text{ ps}$ over 3 GHz
- Simulations of subreflector-feed multipath indicate that -50 dB spec is more restrictive than necessary for path length changes $<$ a few cm.

Source: Brian Corey; Tecspec-Meeting



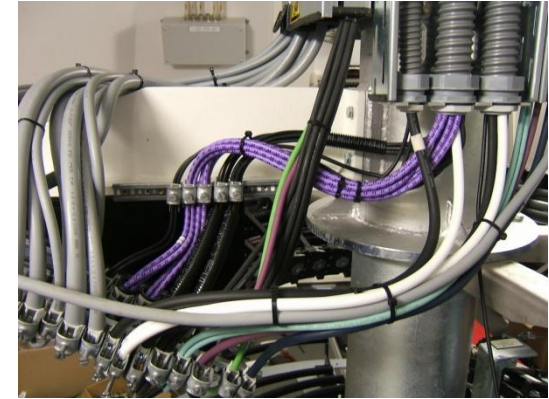
Other TWIN Projects



S/X/Ka-Band Empfänger



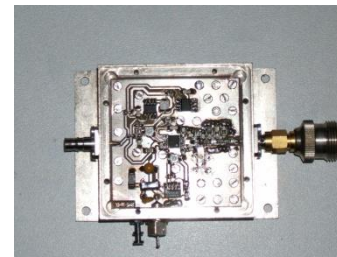
Peltier-Controller für Pcal



Wideband-cable track TTW1



Prototypeboard Monitoring System and A/D Converter

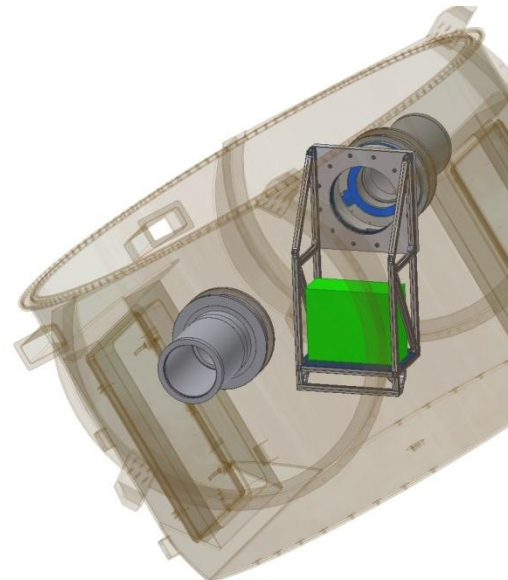
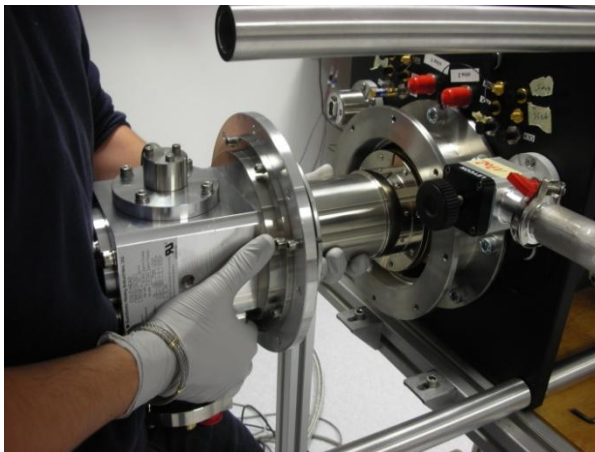
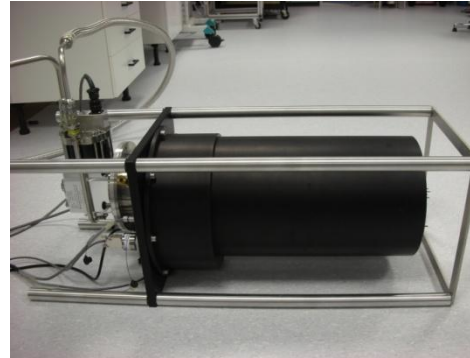


RMS-DC Converter



RF-Equalizer

- Cooled LNA and Feeds are necessary, but how to get the heat out of the cabin?
- How to mount the Helium Compressor in the Elevation cabin?
- How to do a coldhead cylinder maintenance without removing the whole feed?



Other topics: Cable warp and feed blower

- Choosing a cable wrap for 1000 observations a day and a high drive velocity
- Choosing RF-cables for the signal and for the frequency reference
- Choosing fiber cables for a cable wrap and 1000 bend cycles
- How to prevent the feed foils from rain, snow and ice ?



- and so on





- Green Mode Observation
- Loading the deceleration energy back into the power network
- Heat transfer from the electronic devices and servers to warm up the main building
- Good thermal isolation of the main building and the towers



